

Original Article

# PRNGN - PAPR Reduction using Noise Validation and Genetic System on 5G Wireless Network

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**Abstract** - Orthogonal Frequency Division Multiplexing (OFDM) is commonly employed in broadband wireless infrastructures. Due to signal construction discrepancies between the multi-transporter filter, the offset quadrature amplitude modulation (OQAM) signal, and the OFDM signal, the conventional system does not put on to the QAM signal. Undeniably, OFDM has numerous disadvantages, one of which is the high peak-to-average power ratio (PAPR). Connecting and noise reductions are important strategies for this, and the PRNGN network reduces the PAPR reduction problem for the QAM signal using a genetic programme. This approach divides each subset and changes the conventional structure into a single or two-sheet feature search structure, all while using the proposed archetype filter, which has the best PAPR performance of the genetic QAM signal before the PAPR reduction procedure and assists the optimization mode in effectively reducing. This network also suggests a fusion technique for lowering the PAPR of OFDM transmissions. According to simulation findings, the proposed method outperforms the filtering in terms of PAPR reduction and bit error rate (BER) and improves packet delivery.

**Keywords** - Broadband Structure, Filters, OFDM, Optimization, PAPR, Noise Reduction.

## 1. Introduction

Orthogonal Frequency Multiplexing (OFM) is a technique for multiplexing frequencies in an OFDM is a broadband multicarrier modulation system for wireless communication that needs to be updated to be able to use future fifth generation (5G) frequency channels successfully [1-4]. They're similar to video transfer via a wide area network (WAN). During these transactions, many standards must be completed. As a result, this technique can be improved by integrating the filter with multicarrier; also, PAPR reduction can be accelerated. To do so, some strategies are used, such as coding techniques and numerous signal processing. In the proposed PRNGN system, the following methods are used.

### 1.1 Network with 4G Operation

Devices are equipped with Long Term Evolution (LTE) agents such as Serving Gateway (SGW), Mobility Management Entity (MME), Device Manager (DM), evolved node B (eNB), home enode B (HeNB), and User Equipment's (UE) in the 4G communication standard. To access data from the Internet, UE devices are connected to the eNB and HeNB. eNB connects a large number of devices to the network. In addition, eNB allows the handling of channel resources, channel access, and movement monitoring. Furthermore, the OFDM uses numerous user versions of multiple access; the OFDMA radio layer encompasses channel resource regulation, data integration

options, radio connection governing, medium access control layer, physical layer supervisions, and many other features. MME monitors the UE, keeps track of the equipment ID and authentication, and subsequently keeps track of the network environment, among other things. Besides, it is in charge of UE operations and network access. It also provides the equipment with a carrier path. HeNB in LTE supports a small cell company with a Home eNodeB. A HeNB serves the same purpose as an eNB but is designed for indoor communication like a micro device acting as the communal hotspot and has a reduced coverage than a macro eNB. The routing and sharing of user data are the primary functions of the SGW. It is also in charge of internal eNB communication and mobility support between LTE and other networks and supporting data entry points for UEs. It's made to help with radio source supervision, packet transmission according to the quality metrics, inter-cell interference incorporation, and dynamic radio resource supervision.

The beamforming process, including the generation of the beamforming matrix, is maintained in 5G communications by the MAC standard for devices [2]. The number of output beams created is equivalent to the quantity of input data. The number, type, and configuration of the antenna elements in the sorted antenna define their shape. Compared to single beamforming systems, the overall complexity, power consumption, and track are greatly reduced, and the transmission is periodically increased on the



MAC layer of the multiple beamforming technique. OQAM modulation is used to convey data with a good spectral performance and a time-based frequency. OQAM diffuses complex codes in real time and translates the output to a corresponding signal. The quantity of data sets governs which signal system is used. For OQAM symbols, transmission filters are utilized, which are powered by phase modulation using the Inverse Fast Fourier transform (IFFT). Filtering is employed with sample prototype filters to achieve flawless or almost perfect reconstruction. To eradicate multipath transmission channel interference, precision filters are also required [5].

In the second section of this paper, previous ideas and suggestions for this study have been collected from several papers and given as related work. In the third part, the proposed method is given as PAPR reduction using noise validation and a genetic system on a 5G wireless network (PRNGN). Part 4 gives the results of this study and its interpretations, as well as Part 5, concludes the study.

## 2. Related Works

MIMO is a system with many inputs and many outputs. It sends and receives multiple signals simultaneously using multiple antennas on the transmitter and receiver sides [6-7]. Using multiple antennas on the transmitter and receiver side can create a problem caused by multipath dimming. The computer also requires modulation techniques to transmit the signal [8]. OFDM is a special case of multicarrier modulation (MCM). MCM is splitting a signal into multiple signals, converting each new signal to multiple frequency channels, and integrating the data received across multiple channels in the receiver[9].The high PAPR of the OFDM means that if the signal is not to be distorted, many of the components in the transmitter and receiver must have a wide dynamic range[10]. The transmitter's output amplifier must be very linear across a wide range of signal levels. In a wireless system, these amplifiers' cost and power consumption are often important design constraints [11]. The task presented in this paper covers key performance features of the most popular 5G spectrum candidates, as well as practical real-time implementations aimed at field-coded devices that can be programmed under a complex computational analysis and realistic spectrum coherence conditions [12]. These positive characteristics make MIMO-OFDM a promising candidate for high data rate wireless communications. However, the high PAPR of the transmitted signal is a major drawback of the OFDM scheme. Since the MIMO-OFDM system is based on OFDM, it has the same problem. This high PAPR is sensitive to non-linear distortion caused by a high-power amplifier (HPA) [13]. Since simpler devices have limited battery life, it is important to reduce PAPR, which allows for a more efficient high-power amplifier, which can mean longer battery life. In many low-cost applications, the problem of high PAPR may outweigh all the potential advantages of a multicarrier transmission

system [14]. It can have a detrimental effect on battery life in mobile applications. Because simpler devices have limited battery life, it is important to find ways to reduce PAPR, which allows for a small, efficient HPA that can extend battery life [15]. Linear amplifiers are required to avoid radiation and signal distortion outside the bandwidth. Excessive PAPR reduction in multi-user OFDM-MIMO can improve signal quality and ensure appropriate data transfer [16-17]. Suppose the nodes of temporary networks are mobile and have wireless communication to maintain connectivity. In that case, it is known as the mobile ad hoc network. It requires very flexible technology to establish communications in situations that require a fully decentralized network without fixed base stations. Such as battlefields, military applications and other emergency and disaster situations[18]. Predictable, constant noise, the average power of the channel wire reducing PAPR, increases the radio channel and interruption and fading time variation disorders lead to HPA. Unlike the performance of the application. One proven way to mitigate these effects is to use various techniques, such as OFDM [19]. As a highly promising technology, OFDM is used in various communication fields such as digital video broadcasting (DVB), wireless local area network (WLAN) and digital audio broadcasting (DAB). Due to its simplicity, the single-carrier modulation (SCM) scheme is easy to use, but the main problem with SCM is inter-code interference (ICI) [20]. The most popular technique in wireless communications is OFDM. When data is transmitted at high rates through wireless radio channels, the icons may interfere with each other, leading to inter-code interference [21]. OFDM vehicle channels are usually affected by frequency selective dimming, scattering and high PAPR. As a result, linear amplifiers are necessary to avoid radiation and signal distortion outside the bandwidth. Excessive PAPR reduction exposure in multi-user OFDM-MIMO can achieve better signal quality and guarantee adequate data transfer [22]. Space-time encoding with MIMO-OFDM is used in situations where transmission requires data transfer for secure diversity and secure algorithms for data transfer, is there ever a lower PAPR with the MIMO-OFDM system [23]. OFDM is a multicarrier modulation technique in which the bit stream is divided into several orthogonal subcarriers, each modified at a low rate [10]. The module diagram of the OFDM system is described in [1]. Orthogonality is ensured by selecting the appropriate frequency interval between them [24]. MCM is a method of transmitting data by splitting the data into several parallel sub-streams and sending each sub-stream at different frequencies. The reverse FFT acts as an MCM, and the FFT acts as an MCM [25]. Various PAPR reduction methods, such as the active galaxy extension technique, have been proposed to mitigate the occurrence of OFDM signals with large peak power. The ACE program reduces PAPR by repeating time-domain signal clipping and frequency domain constellation point extensions with low bit error rates [26].

### 3. PRNGN Design and Implementation

#### 3.1. Network Creation

This network consists of a set of users  $U(U_1, U_2, \dots, U_n)$ , access points ( $A_p$ ), gateways ( $G_w$ ), and internet servers ( $I_s$ ) in the form of a network graph ( $G$ ). These are designed with 4G and 5G communication standards to communicate. Its users' equipment is outfitted with a MIMO ( $M_{IMO}$ ) antenna for data transfer and reception. The network web server's announcement will be broadcast first. It will initiate the QAM code, which will receive these messages as the gateway, store the server information, and respond to it through a reply gateway message. The server generates the appropriate link ID and updates the gateway ID. Once the access points broadcast their location through a message, the QAM code is run. The random  $R_{nd}$  numbers were used to generate the QAM code ( $QAM_C$ ) between  $R_{nd}(0, QAM_C)$ , as a result, the QAM value is  $QAM_V$  taken from  $R_{nd}$ . The number of QAMs should be calculated as an approximate index at the start of the network based on the size of the lower bound  $L_B$  and the upper bound  $U_B$ . Random numbers from  $R_{nd} = (\frac{L_B}{2}, U_B)$  are considered, and the QAM table is then updated.

The user devices  $U_D$  receive the  $A_p$  notification, which saves the  $A_p$ -ID and its position points. If the user equipment receives several AP-ID announcements, it will use the AP's position points to determine the distance between them and pick the least distant AP according to the location points  $X$  and  $Y$ . If the  $U_D$  receives several  $A_p$  announcements, it will utilize the  $A_p$ 's location coordinates to calculate the distance between them and select the  $A_p$  with the least distance  $D_{ist}$  based on the location points  $X$  and  $Y$  as computed below.

$$D_{ist} = \sqrt{(X_1 - X_2)^2 + (Y_1 - Y_2)^2}$$

User devices then provide a joining report with their location points to the  $A_p$ . When the  $A_p$  receives the joining report from the  $U_D$ , it generates a new user record. Following the establishment of the network connection, each communication node chooses the sub-channel  $S_c$  from the set of channels  $C$  for transmission. The maximum processing signals  $M_s$  are then counted as received symbols  $R_s$ .

$$M_s = \sum S$$

$$R_s = M_s$$

$$S_c = S + R_s$$

The signal's phase is identified based on the odd-even structure of the signal  $S$  during demodulation. Here,  $S$  defines the received actual  $A_l$  and fictional symbols  $F_l$  on the channel.

$$S = A_l + F_l$$

Once the signal is demodulated, the signals are transferred to corresponding filters  $F_T$  to receive a complex signal with  $A_l$  and  $F_l$  portions and the average number of processed signals computed as  $k$ .

$$k = \frac{S + R_s}{2}$$

Here,  $B_N$  which is considered as the block number of data; if the number of  $R_s$  is less than a minimum symbol  $M_B$  then updates the modulated  $m^{th}$  symbol and compute  $S_m$ .

$$S_m = \theta R_s A_l h(m - R_{s1} F_T) + \theta R_s + 1 F_l h\left(m - R_{s1} F_t - \frac{F_T}{2}\right) e^{EI}$$

$$EI = \left[ jk(m - R_{s1} F_T) 2 \frac{\Pi}{F_T} \right]$$

$$j = R_s \bmod B_N$$

$$\theta R_s = 1$$

$$\theta R_{s1} = j$$

The intervention of inter-channel  $C_l$  and intervention of inter-symbol  $C_l$  in  $A_l$  and  $F_l$  configurations are recognized for signal transmission. The predicted power for both  $C_l$  and  $C_s$  will be recognized by the symbol obtained in the sub-channel in  $A_l$  and  $F_l$  sections once the interventions have been identified. In this  $L_s$  is the length of each symbol and  $n^{th}$  symbol on a channel and  $S_s$  is the specific current symbol from the received symbols count  $S_c$ .

$$a1 = S_c (sL_s - S_s) P_R S_c (S_s - R_s L_s) (-\sin R_{s1} - k) \Pi S_s$$

$$a2 = P_R h\left(S_s - R_s L_s - \frac{L_s}{2}\right) \cos(R_{s1} - k) \Pi S_s (-1)$$

$$A_l = a1 - a2$$

$$a3 = S_c \left( sL_s - S_s + \frac{L_s}{2} \right) P_R S_c (S_s - R_s S_s) (\cos(R_{s1} - k) \Pi S_s)$$

$$a4 = P_R S_c \left( S_s - R_s L_s - \frac{L_s}{2} \right) \cos(R_{s1} - k) \Pi S_s (-1)$$

$$F_l = a3 + a4$$

$B_{QRS}B_Q$  of  $A_l$  and  $F_l$  values derived from all data are calculated below, and  $C$  indicates the channel.

$$B_1 = S_c(mC - k)S_c(k - R_s C) \cos \left[ (R_{s1} - k) \left( \frac{2k\pi}{D_s C} \right) + \frac{\pi}{2} \right]$$

$$B_2 = S_c(mC - k)S_c \left( k - R_s C_s - \frac{C_s}{2} \right) \sin \left[ (R_{s1} - k) \left( \frac{2k\pi}{D_s C} \right) + \frac{\pi}{2} \right]$$

$$B_{QRS} = B_1 + S$$

$$B_{QRS1} = B_2 - S$$

Here,  $C_s$  is the sub-channel of the networks. The gaussian random variable is used to distribute the signals independently and evenly, and the predicted signal with interruptions is determined. The predicted value of the signal is used to calculate the signal strength in  $A_l$  and  $F_l$  areas. Filter coefficients related to direct optimization are obtained during the phase transition between channels. The archetype filter strategy is considered to provide low stop band power  $P_r$  for both  $C_l$  and  $C_s$ . The factors listed below are self-governing and evenly scattered interference power calculations.

$$P_r R_s = E_R^2 - 2E_R E_{RS} + E_{RS}^2$$

$$P_r R_{s1} = E_R^2 - 2E_R E_{RS1} + E_{RS1}^2$$

$$E_{RS} = 1 - A_l$$

$$E_{RS1} = 1 - B_l$$

$$E_R = B_1 + B_2$$

During transmission and reception, the power of the  $A_l$  the portion is kept equal to the power of the  $F_l$  portion during the optimization process. Fourier transformation is applied to the filtered signal; the coefficients must be true to achieve an almost flawless reconstruction of the perfect transmission state. The filtered signal's magnitude response is computed to suppress side lobe energy and reduce the side lobe leakage. The stop band region is divided into three segments based on the signal's phase. The minimum energy optimization is applied for the two segments with the threshold power for actual and fictional parts. The objective of the optimization is formulated with the cumulative power value of the considered two regions. Once the optimization is achieved in the filtering process, the signal's peak to average power ratio

is estimated. The  $P_{APR}$  estimation is performed in the discrete time signal with the oversampled signals to the maximum extent. During the consideration, the frame with the overlapping OQAM data blocks is bounded in the length of the frame and the  $P_{APR}$  value is estimated as the  $10\log_{10}$  value of the normalized maximum signal to the ratio of the expected signal frames. In the discrete time signal domain, the disjoint sub-blocks are identified in equal length, and the signal is sequenced into sub-blocks in the  $n$ -symbols with the phase factor vector. By selecting the optimal phase factor vectors, the  $P_{APR}$  is minimized with the maximum value of cumulative vectorized signals. Now the phase rotation is applied to the present data block in the Partial sequence transmission. During this transmission, Genetic computations are applied to exploit the overlapping structure with the upsampling. Here  $R_{SA}$  is the received symbol average, and it is computed based on the received symbol average value  $R_{SV}$  and received symbol last  $R_{SL}$ . Also,  $P$  is the power value.

### Update Bitcounts $B_c$

if  $B_c > 0$  save  $R_s$

$$P = R_{SV}$$

$$R_{SA} = \frac{R_{SV} + R_{SL} + P}{3}$$

The data blocks are partitioned into sub-blocks using the adjacent partition method, and then the minimized  $P_{APR}$  value is computed for the maximized average twice-powered signal.

$$P_{APR} = \frac{R_{SP} + P}{2R_{SA}}$$

When mitigating effects, signal clipping is employed, and the sub chooses the best option. The penalty operators for the data blocks are computed from the distortion value. The traversal will be applied with a signal that matches the initial  $P_{APR}$  value once the penalty operators have been located. The  $P_{APR}$  value for the subgroup is calculated and compared to the initial  $P_{APR}$  value. If it is less than the current average value then the block is assigned with the final set of phase factors. Once the phase factors are assigned, the genetic algorithm is applied to identify the optimal phase factor. The initial population set comprises the available phase factors based on the signals. For each solution in the population, the fitness  $F$  value is computed based on the inverse valued logarithmic of the  $P_{APR}$ . The adaptive probability is computed for the solution in the population, and the parent individuals are identified based on the probability. Here  $\alpha$  and  $\beta$  values are 0.4 and 0.3, considered tunable values.

$$m = \alpha P + \beta R_{SA}$$

$$Fitness = m + (1 - \alpha - \beta)P_{APR}$$

$$F \pm Fitness$$

$$Update F$$

By applying the crossover operations, new offspring are generated, the mutation is applied to the parent individuals based on the adaptive probability, and the new offspring is generated. Here  $F_i$  and  $F_j$  are the crossover parameters.

$$tmp = F_i$$

$$F_i = F_j$$

$$F_j = tmp$$

The fitness value is computed for the newly generated solutions in both crossover and mutation. Here  $i, j$  and  $k$  are the bit values.

$$B = C_H[i][k]$$

$$C_H[i][k] = C_H[j][k]$$

$$C_H[j][k] = B$$

Here  $C_H$  is the chromosomes, and  $B$  is the bits. If the fitness value of the newly generated solution is greater than the fitness value of the parent solution, then the count break is computed. Based on  $P_{APR}$  values, the numerical interval represents a continuous optimal solution with a maximum exercise value. The greatest exercise value is determined from the newly produced solutions, and the related solution is indicated as the best answer. The procedure of determining the best solution for all the solutions in the population and the maximum number gap range is completed. The population for the next iteration is chosen using the roulette wheel method. Repeat ends when the selected optimal solution remains the same for future iterations or when the maximum number of iterations is reached. The selected solution is decoded as the signal strength to broadcast once the optimal solution has been discovered. As inputs, Fast Fourier renovates  $F_{FR}$  calculates the orthogonal signal, number of subcarriers, complex vector, and QAM modulation. On the time and frequency domain signal, the signal value is determined. The exaggerated signal is treated with zero padding in the frequency domain during the approximation of the continuous signal. In congregation charting codes, it's used in the complex vector. The complex vector comprises both actual and fictional portions of the symbol, and the multipath dimmer's tolerance improves

spectral performance. The number of subcarriers is huge, and the independent homogeneity is Gaussian. Thus the amplitude rally distribution is used to determine the independent homogeneity. The probability is determined for this Gaussian probability using the density-dependent variance and the signal amplitude. The entire distribution of the filler is now employed as the  $P_{APR}$  probability value threshold. The amplitude threshold, dependent on the signal phase and average intensity, is used to perform repeated clipping and filtration. Radiation outside the bandwidth is taken into account when designing the filter.

$$while A_l > M_B$$

$$A_l = \frac{A_l}{10}$$

$$while F_l > M_B$$

$$F_l = \frac{F_l}{10}$$

By modulating the signal throughout the reactivation process, this coupling and filtration are used to achieve optimally  $P_{APR}$ . After the connecting mechanism is created, the noise reduction procedure is developed for both the transmitter and the receiver. The input signal is used for reverse  $F_{FR}$  assemblage charting. After calculating the average power, the noise reduction method of  $P_{APR}$  reduction unit is introduced. The noise reduction function is constructed with a normalization factor with a compressed or amplified signal range. The mark function, which takes the phase input into the reverse structure, determines the shape of the noise reduction function. In the linear noise reduction process, the limit value is applied to 1 and linear measurement is used for the parameters, including the output signal's maximum amplitude. The noise reduction function is applied to the signal and the gateway for the high-speed phase and connecting process. And the normalization is applied, and the output scaling factor is fixed based on the shape factor. This function does not replace the signal with small amplitudes, and non-linear contraction is applied to the amplitude adjacent to the threshold. The clipping threshold is recognized as the default factor, which measures both the input and output factors at the same time. Other signals are limited in terms of signal amplitude. The noise reduction procedure included a fixing rate, which unchanged the diminutive signal with variable fixation rates. It limits the large signal to the clipping threshold, which creates a hybrid process of extracting and reducing the noise ratio of the signal. Before utilizing the modulation functions, the  $P_{APR}$  hybridization process reduces value. Assemblage charting is used on the receiver side to recuperate data from the demodulated  $F_{FR}$  signal  $S$ .

### 4. Results and Discussion

Nodes located in this network include an internet server and gateway. Many products operate as access points and user types of equipment. All of these operate on a network of 800 x 800 square meters. All of these operate according to the protocol PRNGN. The table below gives the parameters used in this network.

Table 1. Network Parameters

Parameters	Value
Radio-propagation model	Two Ray Ground
MAC type	Mac802_11
Antenna model	Directional Antenna
Nodes	101
X dimension of topography	800sqm
Y dimension of topography	800sqm
Simulation time	300s
Initial energy in Joules	100
Packet Interval	0.5s
Packet size	900 bytes
Receiving Power	0.01mw
Transmission Power	0.02mw

Table 2. Packet Size Vs PDR

Packet Size	Packet Delivery Ratio (PDR)	
	GABS	PRNGN
900	90.0735	90.1579
910	90.0735	90.1579
920	90.0735	90.1579
930	90.0786	90.1474
940	90.0786	90.1474

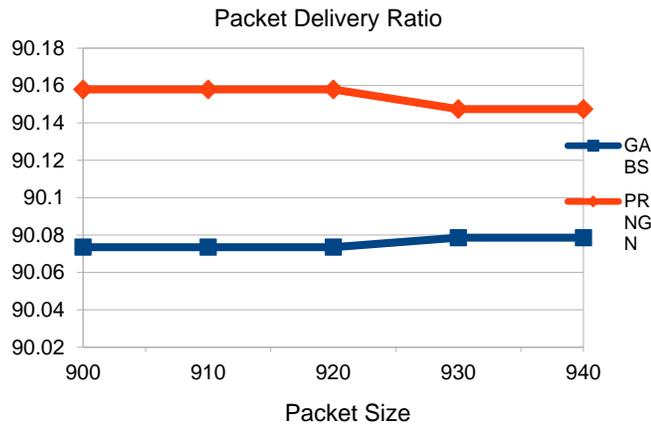


Fig. 1 Packet Size Vs Packet Delivery Ratio

The PDR is a measure that reflects a network's quality. GABPS has delivered a larger percentage of planned packets to the destination than the PRNGN network running on this 5G system. It is because genetic computations are used to construct the PAPR reduction approach alongside the noise reduction method. As a result, information losses are kept to a minimum. As a result, there is a rise in PDR in Figure 1.

Table 3. Packet Size Vs Packet Dropped

Packet Size	Packet Dropped	
	GABS	PRNGN
900	999	991
910	999	991
920	999	991
930	997	989
940	997	989

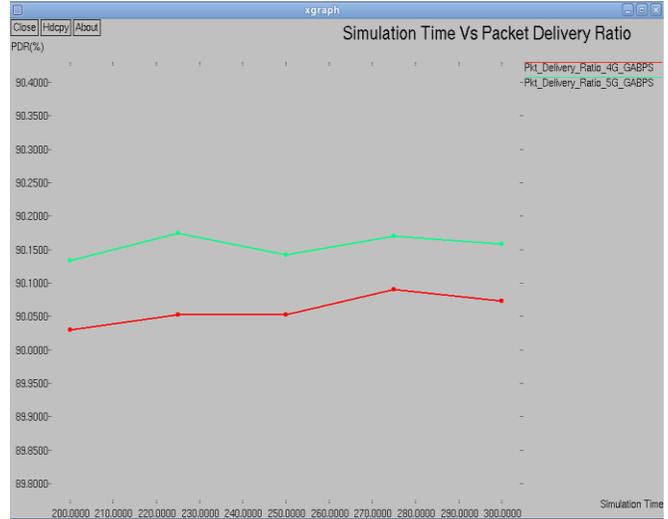


Fig. 2 Packet Size Vs Packet Dropped

If the network functions are uninterrupted, the delivery destination of the packets will increase. Losses occur during network connections according to wireless nature. Reducing them and delivering the data to the destination can be considered an increase in performance. The packet losses are minimized on this PRNGN network, as shown in Figure 2; it tests many features of the network and then sends data, minimizing packet losses.

Table 4. Simulation Time Vs Delay

Simulation Time	Delay	
	GABS	PRNGN
200	0.0146698	0.0130306
225	0.0143336	0.0130347
250	0.0141307	0.0129729
275	0.0139505	0.0128993
300	0.0138221	0.0128545

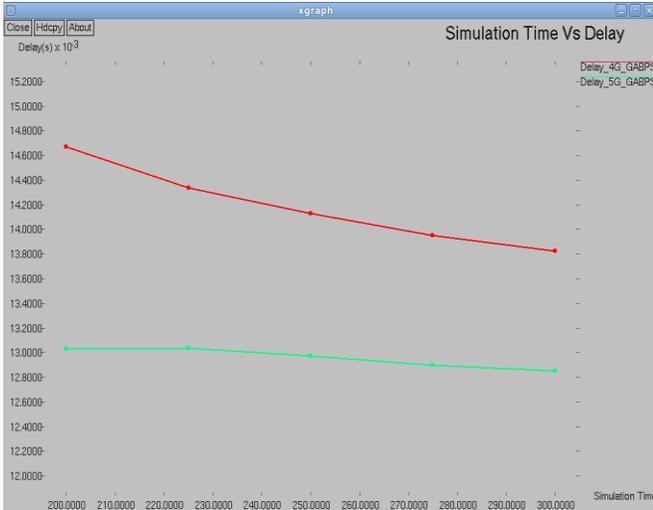


Fig. 3 Simulation Time Vs Delay

Network delay includes channel propagation delay, transmission delay and buffering delay. The protocol running on the network calculates the path along which it will run in the shortest time, along with its important metrics. It will select the path with the least delay in the path selection. It can send more data. In this Figure 3, there is less delay in PRNGN.

Table 5. Simulation Time Vs Goodput

Simulation Time	Goodput	
	GABS	PRNGN
200	501709	564824
225	513477	564646
250	520853	567336
275	527581	570574
300	532479	572560

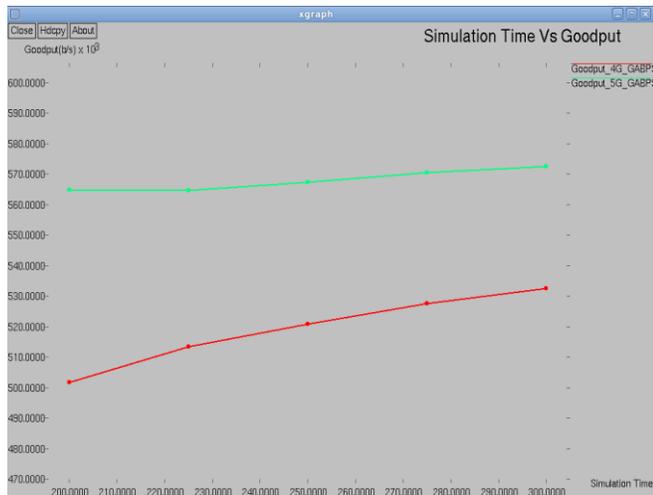


Fig. 4 Simulation Time Vs Goodput

Goodput is the bits count of data packets sent over the network. The more packets exchanged over a network, the destination will receive the more data bits. It can be seen that

a large number of data bits have been received in PRNGN on this network. It can be seen that it is higher than GABS, as shown in Figure.4.

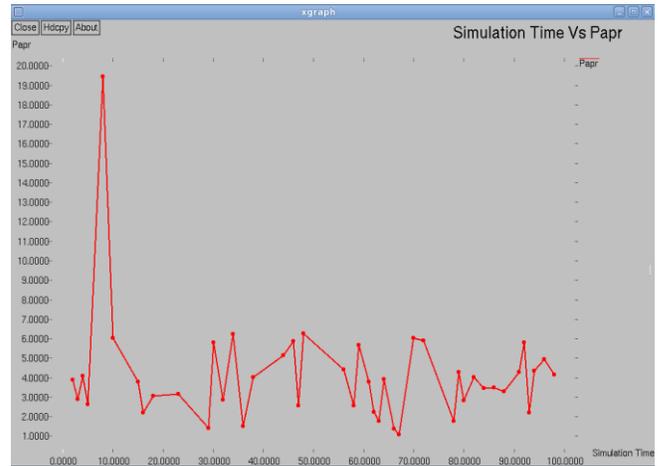


Fig. 5 Simulation Time vs PAPR

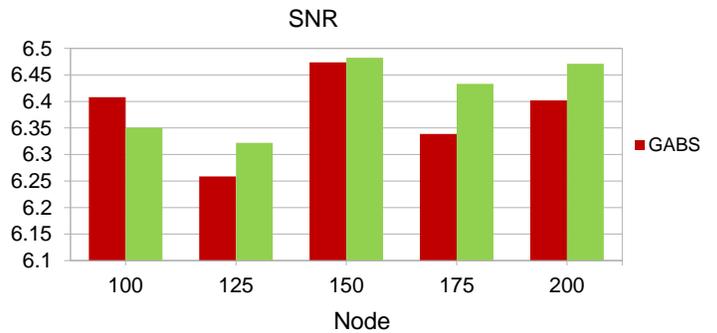


Fig. 6 Node Vs SNR

Table 6. Node Vs SNR

Node	SNR	
	GABS	PRNGN
100	6.40803	6.3505
125	6.25864	6.32205
150	6.47366	6.48223
175	6.33878	6.43326
200	6.40206	6.47105

The signal-to-noise ratio (SNR) explains the signal and noise levels on the observed packets. When the signal exceeds the noise, the quality is good. In this case, the proposed PRNGN produced greater SNR as outputs, as shown in Figure 6, indicating that it will transport packets reliably. The bit error rate (BER) showed the percentage of erroneous bits relative to the total number of received bits in transmission; in this case, Figure 7 depicted the PRNGN's bit error rate. It has the fewest errors in it.

**Table 7. Node Vs BER**

Node	SNR	
	GABS	PRNGN
100	8944	7112
125	16168	6256
150	19464	9376
175	17672	11360
200	19936	9832



**Fig. 7 Node Vs BER**

### 5. Conclusion

The signal is transmitted in this PRNGN network following the data blocks. The transmission filters are applied to the QAM symbols and used with IFFT segment modulation. The sampled filters are used to apply the filtering with the closest reconstruction. After the signal has been demodulated, it is processed through coordinated filters to produce a multifaceted signal containing both actual and

fictional components. Inter-symbol and Inter-channel interference is detected in the actual and fictional structures for signal transmission. Then, determine the potency of the received sign on the sub-channel. The signals are then distributed using a Gaussian arbitrary capricious, and the predicted signal with interferences is calculated. To satisfy the closely perfect reconstruction, the filtered signal is applied to the FT with the coefficients limited to be actual. The PAPR for the signal is estimated once the filtering process has been optimized. A Genetic System is used to leverage the overlapping structure with the upsampling during this transmission. The minimized PAPR value for the maximal average twice powered signal is computed. After the phase factors have been assigned, the genetic computation is used to find the best phase factor. The fitness value is computed for the newly created solutions in both crossover and mutation. The greatest fitness value is computed from the freshly created solutions, and the appropriate solution is identified as the best answer. The number of subcarriers, complex vector and QAM modulation is used to calculate the Fourier applied orthogonal signal. Before conducting the modulation operations, this hybrid approach lowers the PAPR value. The data from the demodulated Fourier signal is retrieved using assemblage remapping at the receiver.

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